

Miniaturized Band Notched UWB Antenna with Improved Fidelity Factor and Pattern Stability

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Abstract. *This article presents a novel miniaturized monopole UWB antenna with enhanced band rejection characteristics at 5.5 GHz. The proposed configuration is derived from octagonal band notched UWB antenna which consists of W-shaped slotted ground plane and an open ended $\lambda_g/4$ hook shaped resonator as a band-notched element. By using proper miniaturization of the octagonal monopole and ground plane, 50% reduction in main structure can be realized without changing the dimension of resonator. Frequency and time domain performance of the proposed miniaturized antenna design is compared with conventional half size and full size structures. The noticeable improvements in terms of band rejection characteristics, radiation pattern stability and fidelity factor confirm the effectiveness of the proposed miniaturization technique. Furthermore, the proposed antenna provides much wider operating bandwidth ($VSWR < 2$) within 3.1–17.2 GHz with the exception around 5–6 GHz.*

Keywords

Miniaturized antenna, ultra wideband (UWB), radiation pattern stability, fidelity factor

1. Introduction

In recent year ultra wide band (UWB) communication system becomes popular among the researchers after the allocation of commercial UWB frequency band of 3.1–10.6 GHz by federal communication commission (FCC) [1]. However, UWB system will cause interference to the existing narrow band systems such as WIMAX system at 3.5 GHz (3.3–3.7 GHz) and WLAN system at 5.5 GHz (5–6 GHz) so that different band rejection element may be employed in the antenna itself to minimize the interference. Different band rejection UWB antenna have been reported earlier [2–8]. Miniaturization of antenna is another important issue in current research work. In previous literature several miniaturization techniques of

UWB antenna have been discussed and their frequency and time domain performance have been analyzed [9–12]. Works on half area reduction of the symmetrical UWB antenna with or without band notch using conventional halving technique have also been reported in most of the literatures which provide straightforward miniaturization technique [12–15]. Conventional half reduction uses the mechanism of simple exploitation of structural symmetry. By proposing a common feeding technique and chopping the symmetrical monopole antenna structure into halves along the magnetic symmetry line of the antenna, better impedance bandwidth can be obtained in UWB antenna [13], [14]. Better time domain performance has also been reported in recent literature for miniaturized UWB antenna [15].

In this paper the frequency and time domain performances of the proposed miniaturized band notched UWB antenna have been discussed. Main focus of this paper is to establish the advantage of proposed reduction mechanism on to the conventional reduction in presence of band notch function. Conventional halving technique gives rise to a perturbation in surface current density and presence of band rejection element leads to enhanced rejection at notched bands. However, stronger notch function will result in more destructive interference which in turn degrades the fidelity. The distinctive feature of proposed miniaturization mechanism is the significant improvement in fidelity factor as well as enhanced band rejection. In proposed technique optimized miniaturization in patch and ground plane also provide 50% area reduction in full size band notched UWB antenna. Moreover the additional advantage of the proposed reduced band notch UWB is the radiation pattern stability improvement over UWB frequency range.

The paper is organized as follows. Initially an octagonal band notched UWB antenna has been designed where quarter wavelength open hook shaped slot resonator has been on the one side of the patch surface which is referred to as full size Non symmetric structure. For the symmetrical Full size band notched structure, a hook shaped resonator pair is embedded to the UWB antenna in next devel-

opment. Further the Symmetrical full structure is reduced to conventional half size structure and the proposed miniaturized structure in next steps. In Sec. 3, the frequency domain and time domain performances of the full size (Non-symmetrical and symmetrical), conventional half size and proposed miniaturized antenna are compared and effectiveness of the proposed structure has been analyzed. A brief comparative study is prepared between the conventional half structure as mentioned in previous literature and the proposed miniaturized structure to illustrate its potential. Finally, this paper is concluded in the last section.

2. Design of Band-notched UWB Antenna and Its Miniaturization

Figure 1(a) illustrates full size printed octagonal monopole antenna with W shaped truncation on ground plane which is constructed on Rogers 5870 substrate. The thickness of substrate is 1.6 mm and the relative permittivity is 2.33. The dimension of full size structure is $36.6 \times 26 \text{ mm}^2$. An open ended quarter wavelength band rejection element has been incorporated to one side of octagonal patch surface to obtain band rejection at 5.5 GHz. The optimized length of hook-shaped open ended slotted resonator can be empirically optimized as

$$X_1 + X_2 + X_3 + X_4 + X_5 - 4t = L_n = \frac{c}{4f_n \sqrt{\epsilon_{\text{reff}}}} \quad (1)$$

where $\epsilon_{\text{reff}} = (\epsilon_r + 1)/2$, c is velocity of light, ϵ_r is relative permittivity of the substrate.

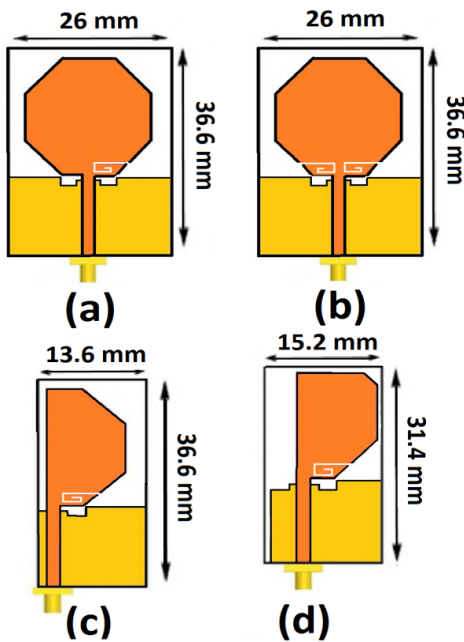


Fig. 1. Antennas: (a) Full size non-symmetrical band notched UWB. (b) Full size symmetrical band notched UWB. (c) Conventional half size band notched UWB. (d) Proposed miniaturized band notched UWB.

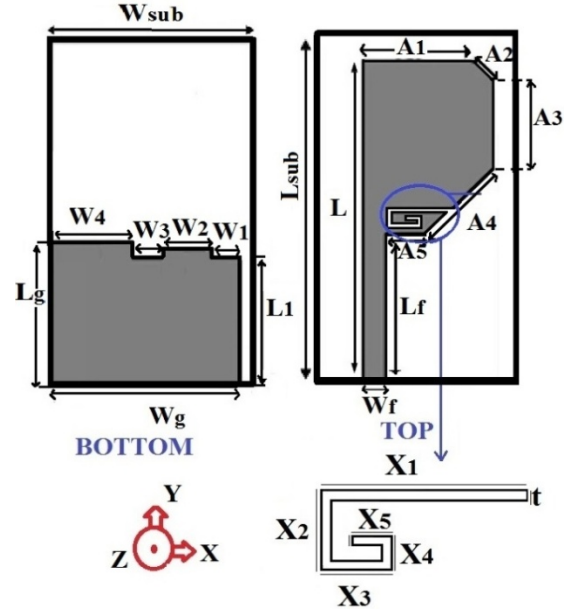


Fig. 2. Geometry and dimension of the proposed miniaturized band notched UWB antenna.

An antenna with one sided resonator is basically a non-symmetric full structure. Insertion of similar resonator pair on the both side of the patch surface makes the band notched UWB antenna a perfect symmetrical full structure as shown in Fig. 1(b). Conventional half size antenna is designed by exploiting the symmetry of radiator and the ground plane. By using the aforementioned conventional halving method, 48% reduction in full size area can be achieved as shown in Fig. 1(c). In this way original full size antenna is simply halved while keeping the patch and ground plane structure symmetrical. However, the remaining part of conventional half size antenna consists of quarter wave length hook-shaped band-notched element. The dimension of conventional half size antenna is $36.6 \times 13.6 \text{ mm}^2$. Since there is no change in the effective length of half size antenna, it does not affect too much on lower operating frequency. The proposed miniaturized structure has been designed by using the proper reduction of length and width of the radiating element and ground plane. The proposed miniaturized band notched UWB antenna has overall dimension of $31.4 \times 15.2 \text{ mm}^2$. In present technique, length of the patch is further reduced while the width of the ground plane is widened towards the truncated part as shown in Fig. 1(d). The length of the ground plane affects impedance matching more significantly at higher frequencies rather than that at lower frequencies.

Actually at higher frequency, current concentration in ground plane becomes stronger; hence multiple truncations on the upper edge of the ground plane contribute towards better impedance matching which in turn suppresses the negative effect of the ground plane at higher frequencies. However, the reduced patch length is sensitive to the low frequency limit and hence the optimized patch length in the proposed antenna has been optimized in such a way that it covers the UWB frequency band. On the basis of the above design consideration, the geometry and prototype of the

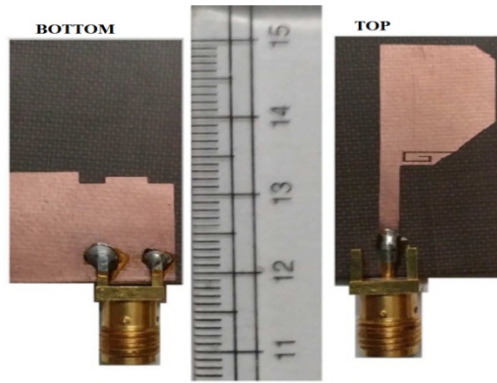


Fig. 3. Photograph of the fabricated proposed miniaturized band notched UWB antenna.

Parameter	Dimension [mm]	Parameter	Dimension [mm]
W_{sub}	15.2	L_2	0.8
L_{sub}	31.4	L_3	0.8
L_g	14	L_4	1.5
W_g	15.2	A_5	3.3
A_1	9.1	L	13.98
A_2	2.8	L_f	1.86
A_3	8.4	W_f	5.36
A_4	8.4	t	0.5
L_1	12.5	W_1	2.5
W_2	4	X_2	1.8
W_3	2.5	X_3	2.8
W_4	8.5	X_4	1
X_1	5.36	X_5	1.3

Tab. 1. Optimized parameter values of the proposed antenna shown in Fig. 2.

proposed miniaturized band notched UWB antenna have been illustrated in Fig. 2 and Fig. 3 correspondingly and the optimized dimension of parameters of the proposed antenna is summarized in Tab. 1.

3. Results and Discussion

3.1 Antenna Miniaturization Effect on Band Rejection Characteristics

Figure 4 illustrates simulated and measured VSWR characteristics of the proposed antenna (predicted by HFSS simulator). The size reduction of full size symmetrical octagonal band-notched UWB antenna yields for two new structures: one is conventional half size band-notched UWB antenna and another is the proposed miniaturized structure. For full size band notched non-symmetrical or symmetrical structure, current flows along current paths of different lengths. Due to the longer radiating current path, the surface current concentration around band notch element does not create much influence at the stop band in full size non-symmetrical structure. However, in full size symmetrical structure, a pair of hook shaped resonators has strong rejection effect which enhances the rejection per-

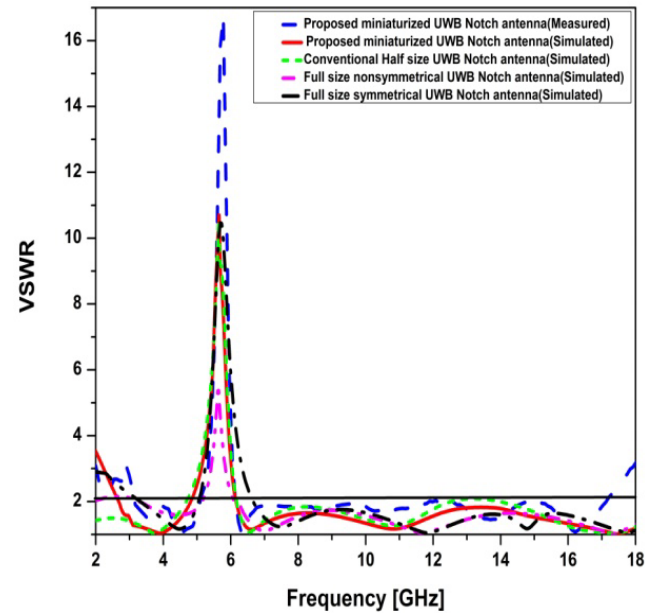


Fig. 4. VSWR characteristics for different structures.

formance almost double. The simulated VSWR performance of full size non-symmetric structure exhibits frequency rejection in the frequency band 5.3–6 GHz with peak VSWR of 4.1 only at 5.5 GHz. On the other hand, full size symmetric structure has wide rejection band (5.2 to 6.5 GHz) with simulated peak VSWR of 10.1. For conventional half size, asymmetrical shape of patch changes the current distribution of the radiating path. Due to the reduction in effective current path, reduced patch surface area accumulates more current and the presence of resonator on that part of the patch surface creates enhanced perturbation which occurs due to the maximum surface current concentration around the band notched element which improves the rejection at 5.5 GHz with simulated peak VSWR of 9. For the proposed miniaturized structure, patch length reduction creates more compact patch size and as a result, deep band notch of simulated VSWR of 10.2 at 5.5 GHz is obtained. The measured VSWR of the fabricated prototype covers the impedance bandwidth 3.1–17.2 GHz (VSWR < 2) with notch band 5.1–6 GHz for WLAN system. The discrepancy in VSWR between simulated and measured results may be attributed to the fabrication tolerance. Quality factor is another focusing issue in band rejection mechanism. Though simulated peak VSWR performance of the symmetrical full structure and the proposed miniaturized structure are same, it is observed that the sharper rejection band of the proposed miniaturized antenna leads to improved quality factor of rejection band.

3.2 Antenna Miniaturization Effect on Radiation Pattern Stability

Radiation properties of the miniaturized monopole antenna are compared with the conventional half size structure, full size symmetric and full size non-symmetric structure. Figure 5 and Figure 6 illustrate the normalized co-polar radiation pattern of four antennas at 4 GHz,

6.5 GHz, 9 GHz and 11 GHz in X-Y plane and X-Z plane correspondingly. The X-Y plane pattern of full size symmetric and non-symmetric structures clearly shows the deep null at $\theta = 90^\circ$ for different frequencies, especially sharp decline at high frequency. The conventional half size design still has the worst angular performance at $\theta = 90^\circ$ but the average null magnitude difference between the conventional half structure and the full size non-symmetric structure is 7.5 dB and approximately 9.2 dB difference with the full size symmetric structure. On the other hand, the patterns of the proposed miniaturized structure have maximum average null difference (12 dB) in comparison to the full size non-symmetric and 14 dB with the symmetric full size structure. Similarly in X-Z plane (as shown in Fig. 6), large turbulence in patterns can be observed in full structures. Patterns of half structure still fluctuate mainly at higher frequencies. But in the proposed miniaturized structure, co-polarized patterns are relatively more stable. In full size structures; the surface current is not uniformly distributed. At higher frequency, full size structures more likely encounter inverse current which affects radiation pattern stability significantly. By using proper miniaturization, the balance between horizontal and vertical surface current of the patch can be obtained. Since width difference between the patch and the ground plane is an important factor for the uniformity of co-polarization patterns, more stable radiation pattern at high frequency can be achieved as the ground plane width is considered to be greater than the effective patch width in the proposed miniaturized antenna. However with increasing ground plane width, impedance matching characteristics are degraded which can be compensated by the multiple etched slots on the upper edge of the ground plane.

Figure 7 illustrates measured co-polarized radiation patterns of the proposed miniaturized structure for azimuth

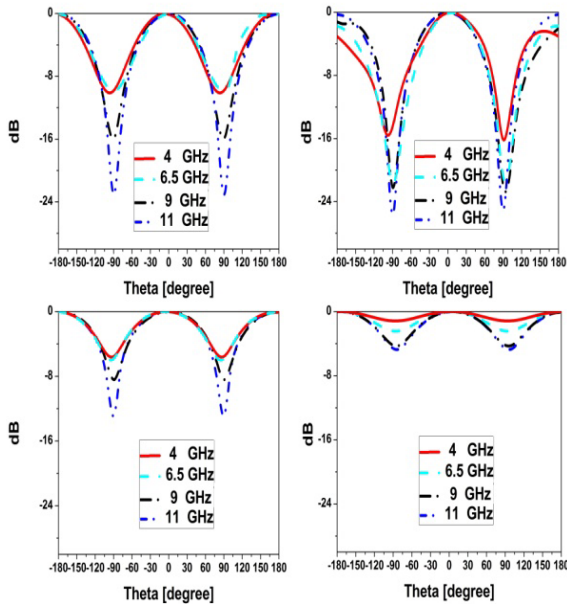


Fig. 5. Simulated elevation (XY) plane radiation pattern for: (a) full size non-symmetric structure, (b) full size symmetric structure, (c) conventional half structure, (d) the proposed miniaturized structure.

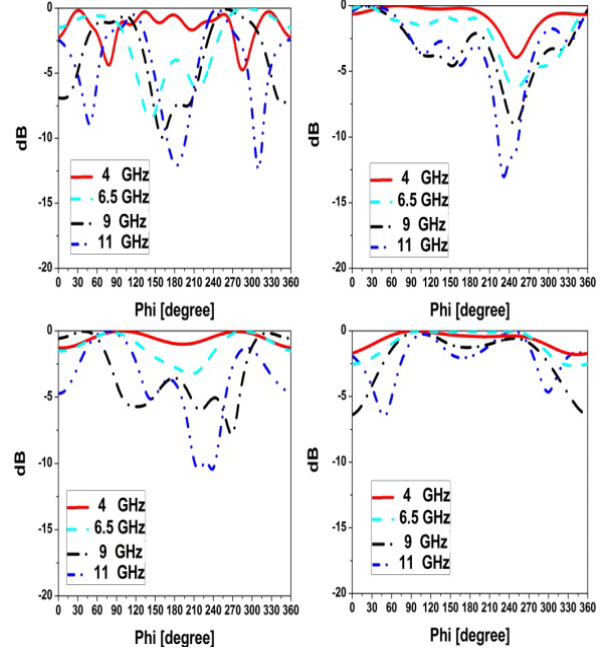


Fig. 6. Simulated azimuth (XZ) plane radiation pattern for: (a) full size non-symmetric structure, (b) full size symmetric structure, (c) conventional half structure, (d) the proposed miniaturized structure.

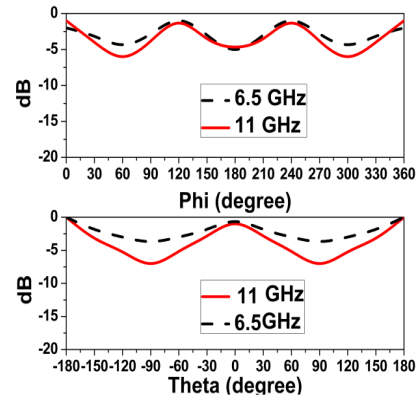


Fig. 7. Measured radiation pattern for the proposed miniaturized antenna for (a) azimuth (XZ) plane, and (b) elevation (XY) plane.

and elevation planes. The measured results of Fig. 7 are in good agreement with the simulated results as shown in Fig. 5(d) and Fig. 6(d).

3.3 Antenna Miniaturization Effect on Fidelity Factor

In order to verify the time domain performance, fidelity factor plays an important role. It is a normalized correlation coefficient between the Gaussian excitation pulse $x(t)$ and the radiated pulse in different direction $Y_{\theta,\varphi}(t)$. Fidelity Factor $FF(\theta, \varphi)$ is given by

$$FF(\theta, \varphi) = \frac{\max}{\tau} \left(\frac{\int X(t) \cdot Y_{\theta,\varphi}(t - \tau) dt}{\sqrt{\int X^2(t) dt} \sqrt{\int Y_{\theta,\varphi}^2(t) dt}} \right). \quad (2)$$

In this paper, the fifth derivative of the Gaussian pulse is assumed as an excitation pulse as shown in Fig. 8. Figure 9 and 10 illustrate the normalized electric field for the different location of virtual probe in elevation (X-Y) plane and azimuth (X-Z) plane for the full size non-symmetric, full size symmetric, conventional half size and proposed miniaturized structures. Pulse preserving capability of the conventional half size structure is worst where the dispersion of pulse can be clearly observed in X-Z plane for the half size structure. Similarly in the symmetrical full size structure, more pronounced ringing effect especially for the lower angle is produced due to the presence of two resonators. On the contrary, the proposed miniaturized structure has a low level of distortion. Similarly in X-Y plane, the intense fluctuation in amplitude can be observed with decreasing value of θ for the conventional half size structure and also the higher time dispersion can be observed in the full size symmetrical structure with increasing value of θ while it becomes smaller in the proposed miniaturized antenna with negligible ringing at trailing edge.

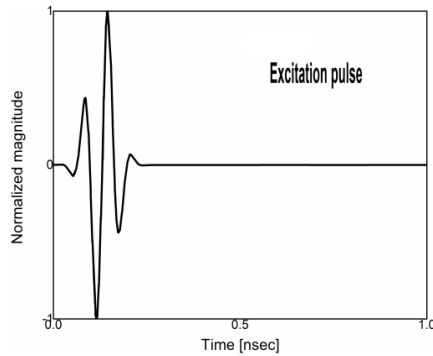


Fig. 8. Gaussian normalized input pulse.

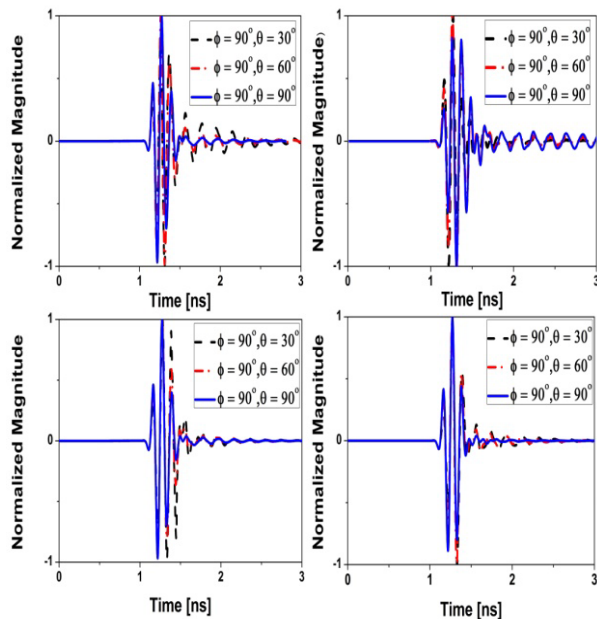


Fig. 9. Received signals (XY-plane) by the virtual probes for (a) non-symmetrical full structure, (b) symmetrical full structure, (c) conventional half structure, (d) proposed miniaturized structure.

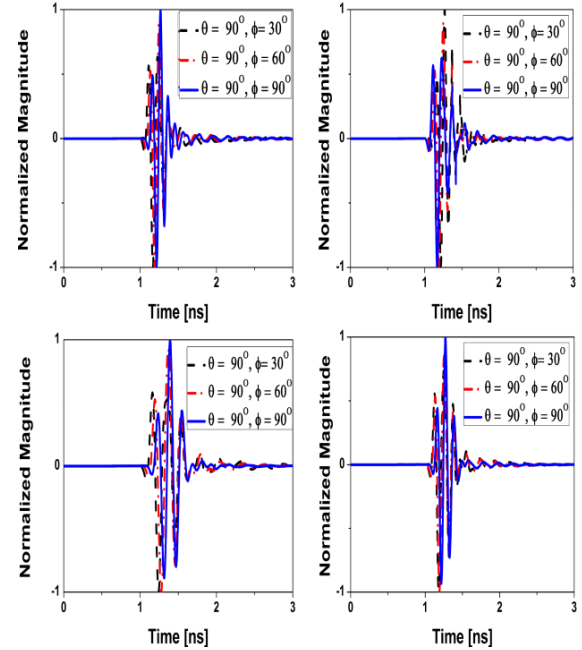


Fig. 10. Received signals (XZ-plane) by the virtual probes for (a) non-symmetrical full structure, (b) symmetrical full structure, (c) conventional half structure, (d) proposed miniaturized structure.

Fidelity factor is plotted for all four structures for elevation (X-Y) and azimuth (X-Z) planes as shown in Fig. 11 and Fig. 12, respectively. The average fidelity factors are summarized in Tab. 2 which reveals the advantage of half size. The hypothesis behind this can be explained as follows. There is a trade-off between quality of notch rejection and time domain performance. Hence poor fidelity factor of the conventional half size band notch structure is mainly due to the negative influence of enhanced notch performance.

Another reason for degraded fidelity factor is the asymmetrical shape of the conventional half size structure. On the other hand, fidelity factor of the symmetrical full

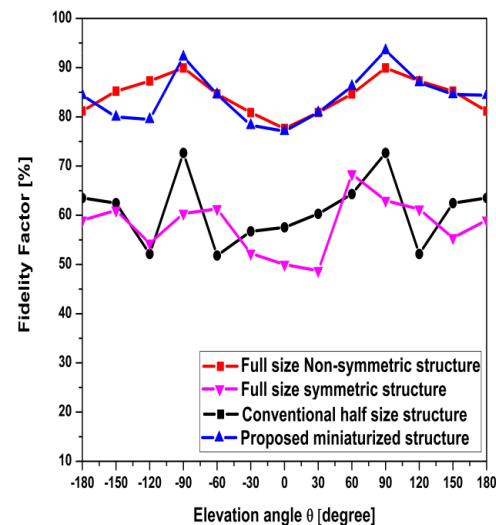


Fig. 11. Fidelity factor in elevation (XY) plane for different values of the elevation angle θ .

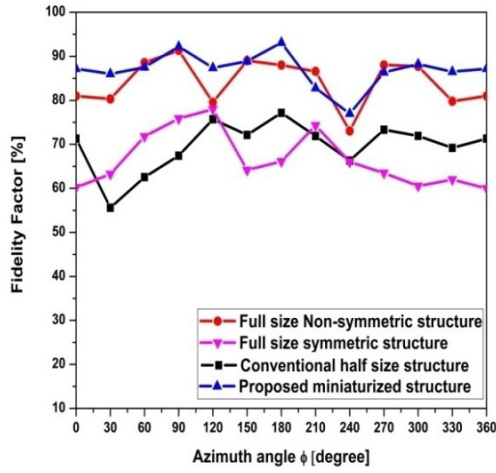


Fig. 12. Fidelity factor in azimuth (XZ) plane for different values the azimuth angle ϕ .

Antenna structure	Fidelity Factor in azimuth plane [%]	Fidelity Factor in elevation plane [%]
Full size non-symmetrical	84.15	84.31
Full size symmetrical	69.33	59.45
Conventional half size	66.56	60.89
Proposed miniaturized	86.18	84.82

Tab. 2. Average fidelity factor of different antenna structure in azimuth and elevation plane.

size structure is also deteriorated due to the presence of band notched element pair. However fidelity factor of the proposed miniaturized structure improves significantly when an abrupt geometric feature of conventional half size antenna is reconfigured.

3.4 Tuning Characteristics of Rejection Band

Figure 13 illustrates the tuning characteristics of the proposed miniaturized antenna around other WLAN narrowband frequencies such as 2.4, 3.6 and 5.2 GHz. Using the arm length variation of open ended hook shaped resonator, notched band can be swept over a wide frequency range and targeted rejection bands can be achieved. As the effective length of the resonator increases, the rejection band shifts towards the lower frequency band.

3.5 Surface Current Distribution

Figure 14 illustrates the simulated surface current distribution of proposed miniaturized antenna at 5.5 GHz where a large current concentration around the hook-shaped resonator is observed.

3.6 Measured Peak Gain

Figure 15 illustrates the measured peak gain of the proposed miniaturized antenna. The average peak gain is 2 dBi throughout the whole operating band except around the rejection band of 5–6 GHz where the peak gain can be as low as −5 dBi.

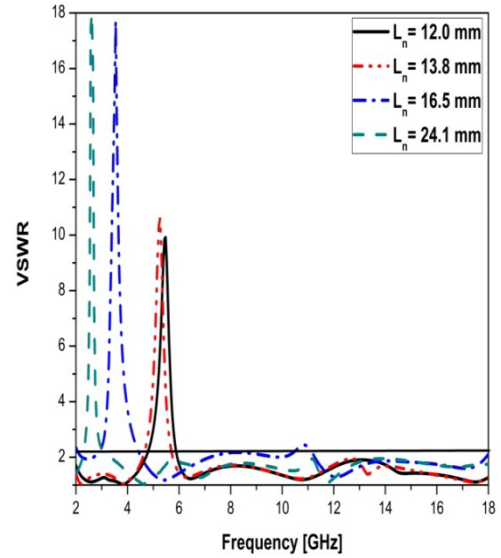


Fig. 13. Simulated VSWR for different values of effective length of the resonator.

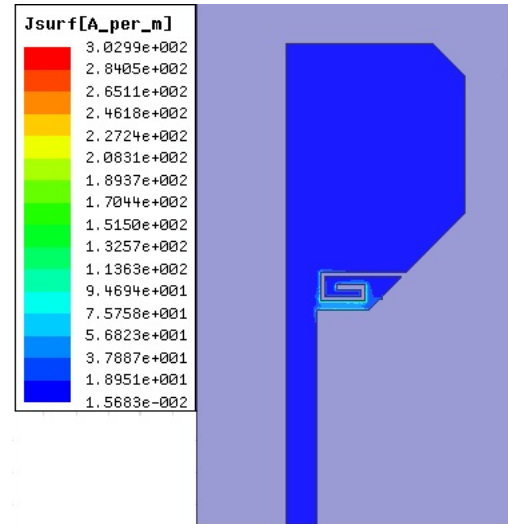


Fig. 14. Simulated result of current distribution at the rejection band frequency of 5.5 GHz.

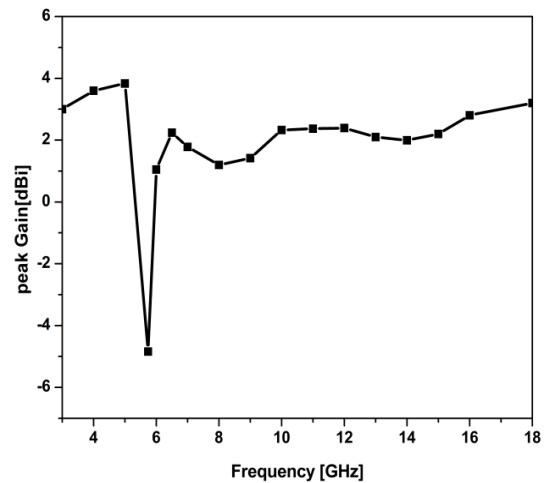


Fig. 15. Measured peak gain of the proposed miniaturized antenna.

Antenna	Conventional half UWB band notched antenna [13]	Proposed miniaturized band notched UWB antenna
Size reduction [%]	50	50
Created notch	Single (fixed at 5.5 GHz)	Single (with tuning facility at 2.4, 3.6, 5.2 and 5.5 GHz)
Quality factor of notched band	3.24	6.17
Radiation pattern stability improvement at elevation plane [%]	84.65	88.39

Tab. 3. Summary of the comparative studies between the conventional half band notched UWB antenna and the proposed miniaturized band notched UWB antenna.

Table 3 summarizes the comparisons of performance investigation and size reduction results of the proposed miniaturized antenna with the conventional half antenna as referred in [13]. In spite of similar percentage area reduction of both, some distinct advantage of the proposed miniaturized antenna is seen over the conventional half antenna irrespective to notched frequency tuning characteristics, rejection band quality factor and radiation pattern stability. On the basis of the above investigation and comparison, it may be concluded that our proposed miniaturization technique is more effective.

4. Conclusion

A miniaturized UWB band notched antenna has been designed and investigated. A novel reduction technique has been adopted for remarkable area reduction of 50% in regular octagonal band notched UWB antenna by keeping the length and the size of the open ended quarter wavelength hook-shaped band notched element is fixed. The proposed structure is compared with another reduced established structure made by conventional halving technique and full structure. The proposed antenna demonstrates significant performance improvement in band rejection, radiation pattern stability and fidelity factor. Moreover, very wide impedance bandwidth (3.1–17.2 GHz) makes the proposed small antenna as a very good candidate for future UWB applications.

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